

MOTOR START-UP CURRENT CONTROL APPARATUS AND METHOD

Cross Reference to Related Applications

This application claims the benefit of United States Provisional Application Number 60/253,216, filed November 27, 2000.

Field Of The Invention

The present invention relates generally to limiting current in a motor during a start-up or a run process. More particularly, the present invention relates to controlling the start-up current and run current of a spindle motor in a disc drive.

Background of the Invention

In the never-ending struggle to produce cheaper, faster, and more reliable computer systems, power consumption has taken on a significant role. Specifically, power to operate the disc drive is supplied by the computer; therefore, the voltage and current demands of the disc drive must meet the specifications that are imposed by the computer power supply capabilities. As development of computer systems have progressed, the amount of power available to disc drives has decreased or remained the same while the speed (revolutions-per-minute) and capacity of the disc drive has increased. As disc drives achieve faster revolutions-per-minute (RPMs), the amount of power needed to achieve the faster RPMs could also increase. Thus, there is a need to control and limit the amount of power used by the disc drive.

One way to accomplish controlling and limiting the power consumed by the disc drive is to monitor and limit the peak motor current through the spindle motor of the disc drive. An example of this is described in U.S. Patent No. 5,216,343 (Genheimer et al.).

While this method is effective in controlling the peak current through the spindle motor, it does not uniformly control the current drawn from the power supply. In fact, significant peaks in the current drawn from the power supply exist at multiple stages during the motor spin-up. These peaks can cause instability, dropouts, and power faults in some power supply systems.

Due to the highly competitive nature of the disc drive industry, it is necessary to produce disc drives which meet customer requirements at the lowest possible price. One method to accomplish this is to lower costs by reducing the number of components needed within the disc drive. This highlights another problem associated with controlling the peak current through the spindle motor: the control mechanisms require additional electronics and hardware components to be added to the disc drive.

The present invention provides a solution to these and other problems, and offers other advantages over the prior art.

Summary of the Invention

The present invention relates to controlling the start-up current and run current of a spindle motor.

In accordance with one embodiment of the invention, a method for controlling the start-up power of a motor is provided in which the amount of current through a spindle motor is monitored. In another embodiment of the invention, a motor start-up sequence is disabled if the correct conditions

materialize. Another embodiment includes using a preprogrammed start-up disc profile as a voltage reference from a digital-to-analog converter.

In yet another embodiment of the present invention, a method is provided for controlling the current drawn from a power supply. Another embodiment of the invention allows the motor drivers to be disabled.

In accordance with another embodiment of the invention, a method for controlling the run power of a motor is provided in which the amount of current through a spindle motor is monitored. In another embodiment of the invention, a motor run sequence is disabled if the correct conditions materialize. Another embodiment includes using a preprogrammed run disc profile as a voltage reference from a digital-to-analog converter.

In another embodiment of the present invention, the invention also can be implemented as a data storage device itself.

These and various other features as well as advantages which characterize the present invention will be apparent upon reading of the following detailed description and review of the associated drawings.

Brief Description of the Drawings

FIG. 1 is a plan view of a disc drive incorporating a preferred embodiment of the present invention showing the primary internal components.

FIG. 2 provides a functional block diagram of the disc drive of **FIG. 1**.

FIG. 3 is a diagram of a preferred embodiment current control apparatus which can be used in the system of **FIG. 1**.

FIG. 4 is a graphical representation of a spindle motor start-up profile which can be used in the apparatus of **FIG. 2**.

FIG. 5 is a flowchart detailing a preferred embodiment current control method.

Detailed Description

A disc drive **100** constructed in accordance with a preferred embodiment of the present invention is shown in FIG. 1. The disc drive **100** includes a base **102** to which various components of the disc drive **100** are mounted. A top cover **104**, shown partially cut away, cooperates with the base **102** to form an internal, sealed environment for the disc drive in a conventional manner. The components include a spindle motor **106** that rotates one or more discs **108** at a constant high speed. Information is written to and read from tracks on the discs **108** through the use of an actuator assembly **110**, which rotates during a seek operation about a bearing shaft assembly **112** positioned adjacent the discs **108**. The actuator assembly **110** includes a plurality of actuator arms **114** which extend towards the discs **108**, with one or more flexures **116** extending from each of the actuator arms **114**. Mounted at the distal end of each of the flexures **116** is a head **118** that includes an air bearing slider enabling the head **118** to fly in close proximity above the corresponding surface of the associated disc **108**.

During a seek operation, the track position of the heads **118** is controlled through the use of a voice coil motor (VCM) **124**, which typically includes a coil **126** attached to the actuator assembly **110**, as well as one or more permanent magnets **128** which establish a magnetic field in which the coil **126** is immersed. The controlled application of current to the coil **126** causes magnetic interaction between the permanent magnets **128** and the coil **126** so that the coil **126** moves in accordance with the well-known Lorentz relationship. As the coil **126** moves, the actuator assembly **110** pivots about

the bearing shaft assembly **112**, and the heads **118** are caused to move across the surfaces of the discs **108**.

The spindle motor **106** is typically de-energized when the disc drive **100** is not in use for extended periods of time. The heads **118** are moved over park zones (not shown) near the inner diameter of the discs **108** when the drive motor is de-energized. The heads **118** are secured over the park zones (not shown) through the use of an actuator latch arrangement, which prevents inadvertent rotation of the actuator assembly **110** when the heads are parked. The heads **118** may alternatively be parked on ramps (not shown) at the outer diameter of the discs **108** when the drive motor is de-energized.

A flex assembly **130** provides the requisite electrical connection paths for the actuator assembly **110** while allowing pivotal movement of the actuator assembly **110** during operation. The flex assembly includes a printed circuit board **132** to which head wires (not shown) are connected; the head wires being routed along the actuator arms **114** and the flexures **116** to the heads **118**. The printed circuit board **132** typically includes circuitry for controlling the write currents applied to the heads **118** during a write operation and a preamplifier for amplifying read signals generated by the heads **118** during a read operation. The flex assembly terminates at a flex bracket **134** for communication through the base deck **102** to a disc drive printed circuit board (not shown) mounted to the bottom side of the disc drive **100**.

FIG. 2 provides a functional block diagram of the disc drive **100**. Data and host commands are provided from a host device to the disc drive **200** using interface circuitry **218** in conjunction with a top level control processor **220**. Data is transferred between the discs **208** and the host device using the interface circuitry **218**, a read/write channel **222**, a preamplifier circuit **224**, and a head **212**.

Head positional control is provided by a closed-loop servo circuit **226** comprising demodulation circuitry **228**, a servo processor **230** (preferably comprising a digital signal processor, or DSP) and motor control circuitry **232**. The motor control circuitry **232** applies drive currents to the actuator coil **214** to rotate the actuator assembly **110**. The motor control circuitry **232** further applies drive signals to the spindle motor **206** to rotate the discs **208**.

FIG. 3 provides a functional system diagram of relevant portions of the motor control circuitry **232** of **FIG. 2**. **FIG. 3** is also a system diagram of the preferred embodiment current control circuit **300**. The current control circuit **300** includes a power supply **302** which is provided by a host computer system (not shown). The power supply **302** powers a spindle motor **304**. The voltage across current sensing resistor **306** is measured when the calibrating switch **314** is not enabled. Calibrating switch **314** may be enabled during the power-up procedure of a drive.

In the preferred embodiment, the calibrating switch **314** provides a calibration reference signal for a 'once at power-up' calibration of the electronics from the digital-to-analog converter (DAC) **310** through comparator **318**. The 'once at power-up' calibration is achieved by applying a precision reference voltage (not shown) through the initiating switch **314**. This allows calibration of the DAC **310** by using a command from the DAC **310** to adjust the DAC **310** reference voltage input into the comparator **318**. Then a comparison can be made between the DAC **310** voltage input and the precision reference voltage by analyzing the comparator **318** output trip threshold. This calibration procedure can be used to eliminate offsets in the current control circuitry. Specifically, this calibration method can be used to measure offsets in the comparator **318**, the gain multiplier **308**, and the DAC **310**.

When calibrating switch **314** is not enabled, gain multiplier **308** multiplies the voltage from the current sensing resistor **306** by a predetermined gain. When switch **316** is closed, the output of the gain multiplier **308** is then applied to capacitor **312**, integrating the output of the gain multiplier **308**. The integrating capacitor **312** produces a voltage at its terminals proportional to the total current applied to the motor **304**. When this voltage from the integrating capacitor **312** reaches a level equal to or greater than the voltage reference set by the DAC **310**, one-shot comparator **318** will fire a finite, programmable duration pulse which will disable motor drivers **320** and reset switch **316**. After being triggered, reset switch **316** discharges the capacitor **312** before the cycle repeats. After the finite programmable duration pulse time expires, the motor drivers are re-enabled and the process repeats.

The charge in the capacitor **312** is measured by at least one input of voltage comparator **318**. DAC **310** supplies a reference voltage to another input of the voltage comparator **318**. The reference voltage from the DAC **310** is determined by a preprogrammed velocity dependent reference profile **400** as illustrated in **FIG. 4**. The reference voltage from the DAC **310** does not need to be set to a specific reference profile, the reference voltage from the DAC **310** may be set to any value, such as a constant value or a time-dependent value.

A velocity dependent reference profile **400** may be stored in memory (not shown) accessible by the DSP **230**. The motor control circuitry **232** includes circuitry (not shown) that measures the velocity of the discs **108**. DSP **230** transmits a data value to the DAC **310** where the data value corresponds to the preprogrammed velocity dependent reference profile **400**.

The voltage comparator **318** is preferably a one-shot comparator which starts a programmable timer/counter (not shown) to disable the motor drivers

320 when the voltage at the capacitor's **312** terminals exceeds the voltage provided by the DAC **310**. The motor drivers **320** are then disable for a programmed amount of time before the cycle repeats. The cycle is repeated by re-engaging the motor drivers **320**.

FIG. 5 provides a flow chart for monitoring the start current or run current of the spindle motor **304** and the current drawn from the power supply **302**, generally illustrative of the steps carried out in accordance with preferred embodiments. The current controlling routine **500** is preferably executed each time the drive is brought from a deactivated to an operationally ready state. The current controlling routine **500** may also be implemented while the drive is in a run process.

The routine starts at step **502** to initialize the start-up procedure for the spindle motor. This preferably includes providing power to the spindle motor from a rest state. The preferred embodiment also includes calibrating the circuitry and determining the initial setting for the reference profile **400**.

After start-up procedure is initialized during step **502**, the voltage of the current sensing resistor **306** is then measured in step **504**. The voltage from step **504** is then multiplied by a gain multiplier in step **505**. Next, the multiplied voltage from step **505** is integrated in step **506**. The integration is preferably done using a capacitor **312** at the output of the gain multiplier **308** and at the input of the voltage comparator **318**.

After the multiplied voltage is integrated, step **508** compares the integrated voltage to a reference voltage value. This is preferably done using a voltage comparator **218** and a DAC **310** with a preprogrammed reference voltage. If the integrated voltage value is less than the reference voltage value, then the routine will repeat back step **502**. If the integrated voltage value is greater than or equal to the reference voltage value then the process proceeds to step **510**.

In step **510**, the process disables the motor drivers. This is preferably done by sending a disable signal to the spindle driver control logic (not shown) in the motor control circuitry **232**. After step **510** is complete, the routine proceeds with timing delay **512**. The routine counts the amount of time on step **512** until a preprogrammed time has passed. When the preprogrammed time has passed the process enables the motor drivers in step **514**. Re-enabling the motor drivers, step **514**, may also include re-enabling voltage to the motor. After the drivers are enabled, the process repeats back to the measure voltage step **504**. In the preferred embodiment, the measure voltage step **504** also includes monitoring the motor velocity and adjusting the DAC **310** reference voltage according to a velocity dependent reference profile, such as **400**.

One advantage of the current control system **300** over other designs is that the amount of voltage and current drawn from the power supply is limited. Controlling the power supply in this manner reduces high frequency current spikes seen on the power supply due to motor commutation switching during spindle motor start-up. Another of the advantages of the current control system **300** over other types of designs is the use of the DAC **310** provides a programmable means for adjusting the reference voltage without having to place additional components on a circuit board. Using a DAC **310** provides the ability to have variations as to what type of reference voltage is used. The reference voltage may be determined by values such as constants, time dependent values, velocity dependent values, or any other value.

Alternately characterized, a first contemplated embodiment of the present invention includes a method in which controlling the start-up power of a motor is monitored. The method comprises steps of initially applying power to a spindle motor to engage a start-up sequence (such as step **502**). Thereafter, the amount of current applied to the spindle motor during the

start-up sequence is monitored (such as step 504). Preferably, the method also includes obtaining a control voltage proportional to the motor voltage and disabling the start-up sequence if the control voltage exceeds a predetermined voltage threshold (such as steps 505 and 508). Preferably, the voltage threshold (such as step 508) is obtained from a digital-to-analog converter (such as 310). In preferred embodiments, the voltage threshold corresponds to a preprogrammed start-up disc profile (such as 400). The method further preferably comprises obtaining the control voltage by integrating (such as step 506) a voltage across a current sensing resistor (such as 306). Further, the method preferably comprises a calibration procedure initialized by a calibration switch (such as 314).

A second contemplated embodiment of the present invention includes a method for controlling the current drawn from a power supply in a computer system. The method comprises steps of initially applying power to a spindle motor to engage a start-up sequence (such as step 502). Thereafter, the amount of voltage applied to the spindle motor during the start-up sequence is monitored (such as step 504). Preferably, the method also includes obtaining a control voltage proportional to the motor voltage and disabling the start-up sequence if the control voltage exceeds a predetermined voltage threshold (such as steps 505 and 508). Preferably, the voltage threshold is obtained from a digital-to-analog converter (such as 310). In preferred embodiments, the voltage threshold corresponds to a preprogrammed start-up disc profile (such as 400). The method further preferably comprises obtaining the control voltage by integrating (such as step 506) a voltage across a current sensing resistor (such as 306). Further, the method preferably comprises a calibration procedure initialized by a calibration switch (such as 314).

A third contemplated embodiment is a data storage device including at least one spindle motor, a power supply electrically coupled to the spindle motor (such as 302), and a spindle motor controller (such as 232). The spindle motor controller (such as 232) is operatively coupled to the spindle motor and the power supply so as to measure and limit an amount of power from the power supply that is utilized by the spindle motor during a spindle motor start-up sequence (such as 500) or a spindle motor run sequence. Optionally the data storage device includes a driver control function programmed into the motor controller so that a spindle motor driver (such as 320) may be disabled for a fixed period of time.

Preferably, the data storage device includes a data storage device controller (such as 220) which is operably coupled to the spindle motor controller (such as 232), the data storage device controller having the functionality to initiate or deactivate the spindle motor start-up sequence.

In preferred embodiments, the driver control function disables the spindle motor drivers when a signal proportional to a current through the spindle motor exceeds a predetermined threshold. Preferably, the predetermined threshold is a programmable voltage from a digital-to-analog converter (such as 310).

Another contemplated embodiment of the present invention includes a method in which controlling the run power of a motor is monitored. The method comprises steps limiting the amount of current applied to the spindle motor during the run sequence (such as step 504). Preferably, the method also includes obtaining a control voltage proportional to the motor voltage and disabling the run sequence if the control voltage exceeds a predetermined voltage threshold (such as steps 505 and 508). Preferably, the voltage threshold (such as step 508) is obtained from a digital-to-analog converter (such as 310). In preferred embodiments, the voltage threshold corresponds to

a preprogrammed start-up disc profile (such as 400). The method further preferably comprises obtaining the control voltage by integrating (such as step 506) a voltage across a current sensing resistor (such as 306).

It is to be understood that even though numerous characteristics and advantages of various embodiments of the present invention have been set forth in the foregoing description, together with details of the structure and function of various embodiments of the invention, this disclosure is illustrative only, and changes may be made in detail, especially in matters of structure and arrangement of parts within the principles of the present invention to the full extent indicated by the broad general meaning of the terms in which the appended claims are expressed. For example, the particular elements may vary depending on the particular application for the current control circuit and method while maintaining substantially the same functionality without departing from the scope and spirit of the present invention.

In addition, although the preferred embodiment described herein is directed to a current control apparatus and method for a disc drive system, it will be appreciated by those skilled in the art that the teachings of the present invention can be applied to other systems, like floppy drives, CD-ROMs, and DVD players, or any other system employing a motor without departing from the scope and spirit of the present invention.